

Model for Nonlinear Interference Noise in Raman-amplified WDM Systems

Presenter: *Marco Santagiustina*



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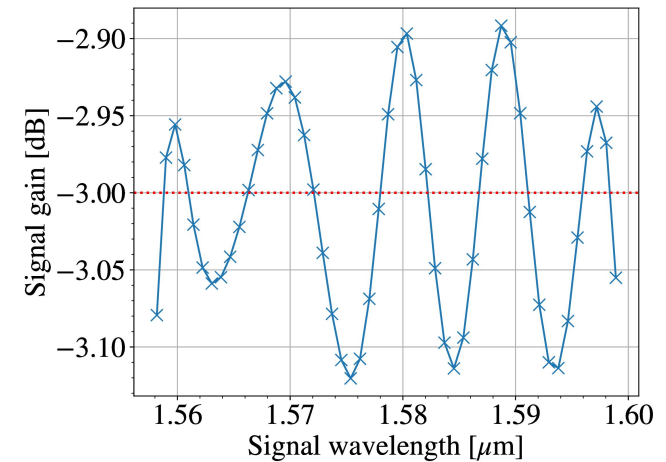
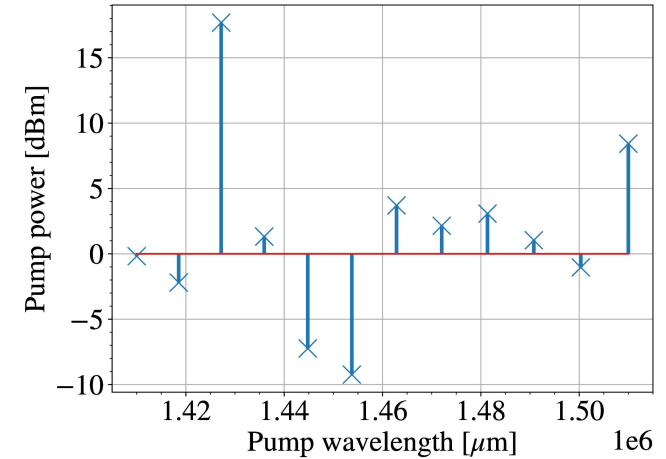
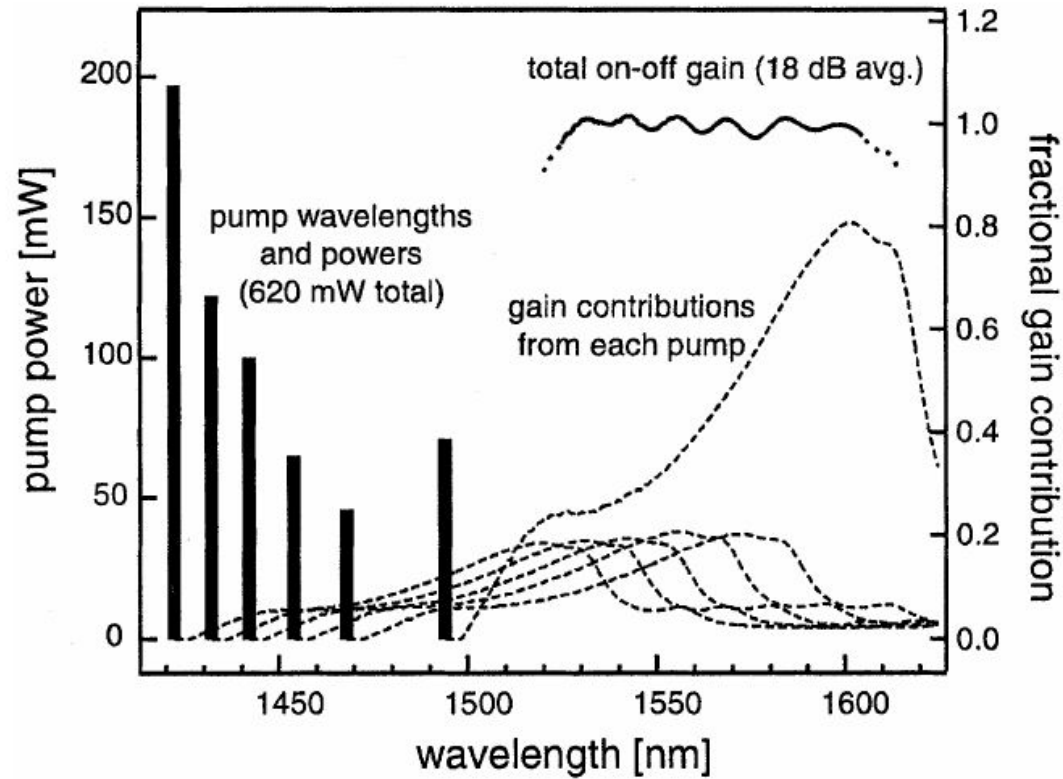
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Outline

- Brief **introduction** to Raman-amplified WDM systems and **motivation** on the study of NonLinear Interference Noise (NLIN).
- Illustration of the **extension** of the existing **model** for NLIN, for these systems.
- Description of the consequences ,in terms of noise performances, by using **numerical simulations**.

Raman-amplified WDM systems: equalized amplification



- [1] J. Bromage, *JLT* (2004).
- [2] G. Marcon et al., *JLT* (2021).

Nonlinear Interference Noise

The model by Dar and Mecozzi [3] assumes **perfect amplification**, thus using NLSE of the form

$$\frac{\partial}{\partial z} u = -i \frac{\beta_2}{2} \frac{\partial^2}{\partial t^2} u + i \gamma |u|^2 u$$

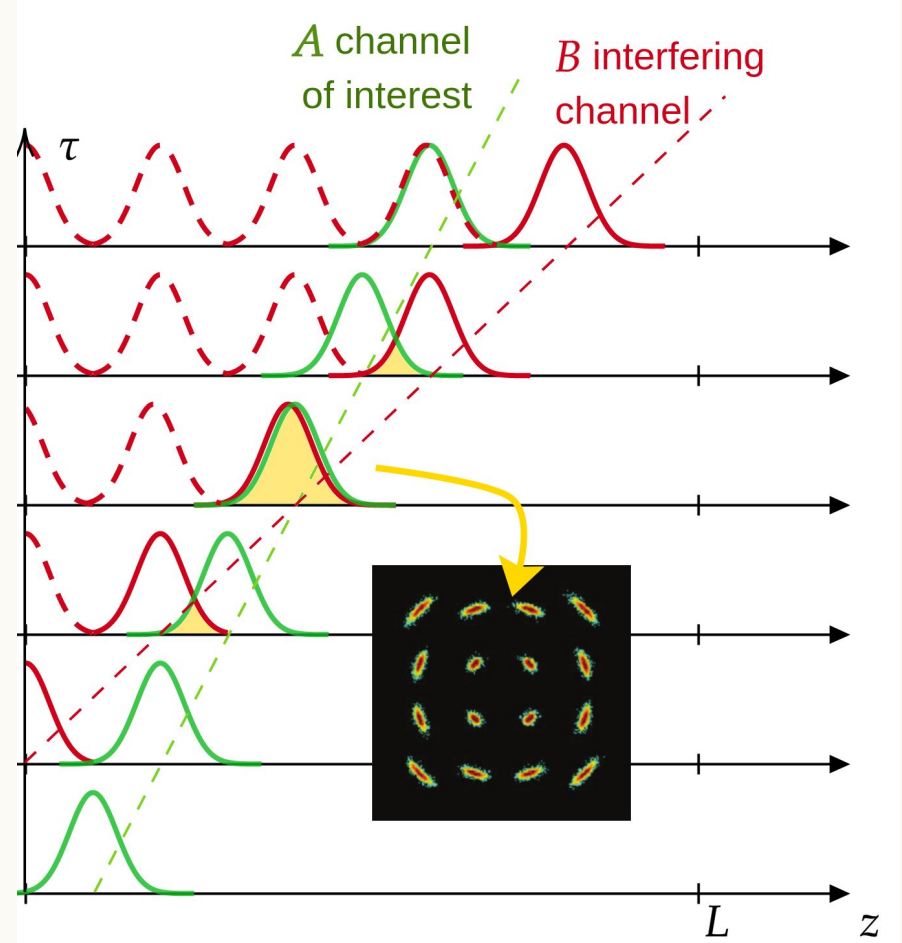
and obtains **phase noise variance** using first order perturbation theory,

$$\Delta a_0 = i 2 \gamma a_0 \sum_m |b_m|^2 X_{0,m,m},$$

$$X_{0,m,m} = \int_0^L dz \int_{-\infty}^{+\infty} dt |g^{(0)}(z, t)|^2 |g^{(0)}(z, t - mT - \beta_2 \Omega z)|^2.$$

Analytical solution is found by assuming to have **high dispersion**, with the approximation

$$X_{0,m,m} \approx \frac{1}{\beta_2 \Omega}.$$



[3] R. Dar et al., *Opt. Express* (2013).

Computation of total phase noise

The phase noise variance is

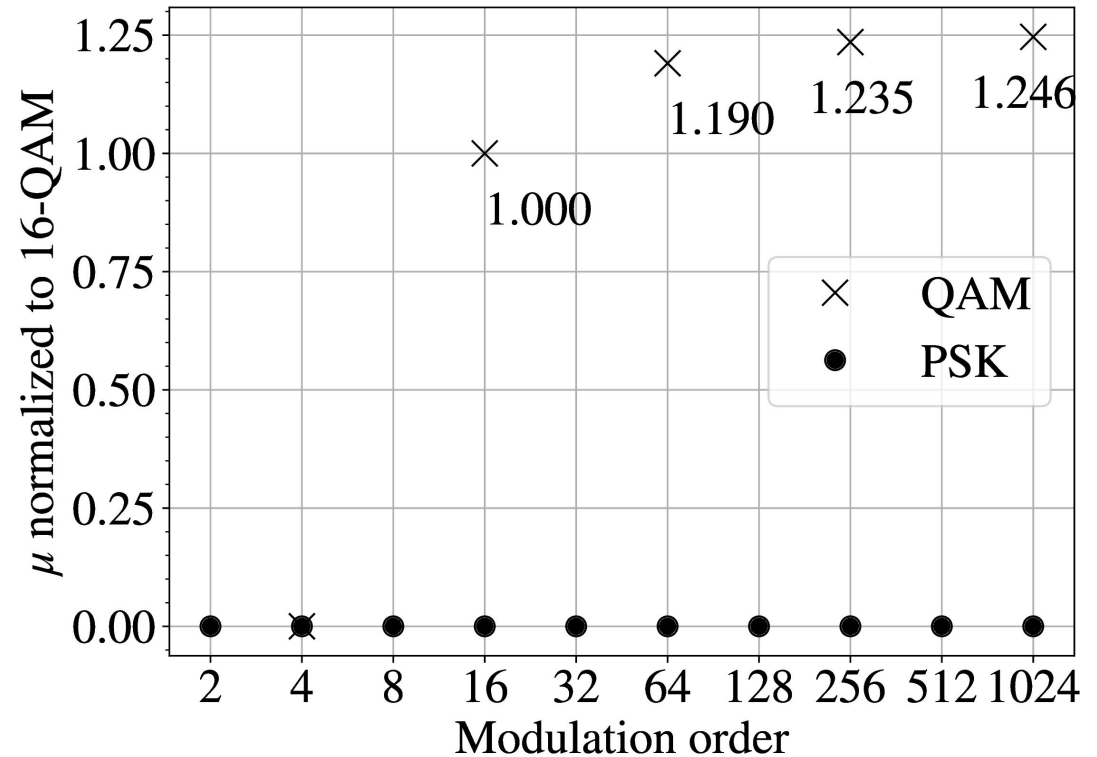
$$\Delta\theta^2 = 4\gamma^2 \underbrace{\left(\mathbb{E} [|b_0|^4] - \mathbb{E} [|b_0|^2]^2 \right)}_{\text{modulation format}} \underbrace{\sum_m X_{0,m,m}^2}_m_{\text{pulse overlap}}$$

and the expression can be decomposed as

$$\Delta\theta^2 = 4\gamma^2 (P_B T)^2 \mu \sum_m X_{0,m,m}^2$$

$$\mu = \left(\frac{\mathbb{E} [|b_0|^4]}{\mathbb{E} [|b_0|^2]^2} - 1 \right),$$

separating impacts of *pulse overlaps*, *input power* and *baud rate*, and *modulation format*.

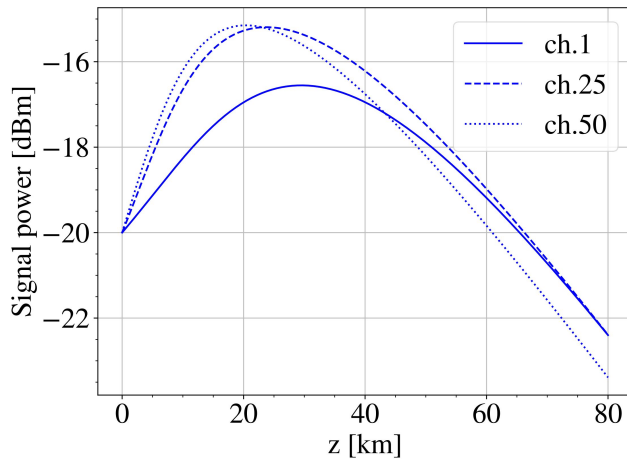


Waves power evolution over fiber length

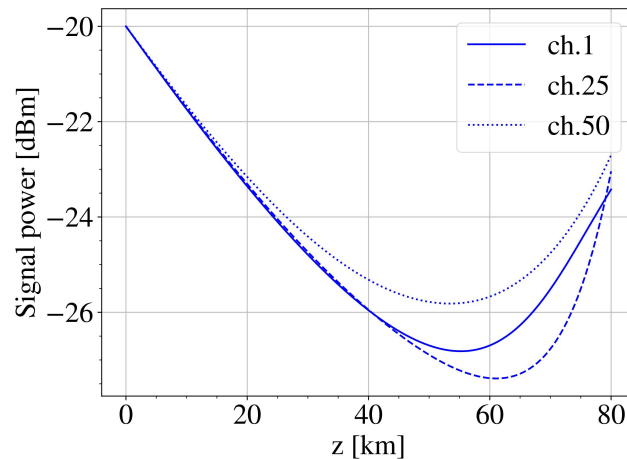
Power evolution equations accounting for **Raman gain** function C_R , and **attenuation**, are the following:

$$\pm \frac{dP_i^\pm}{dz} = -\alpha_s P_i^\pm + \left[\sum_{j \neq i} C'_{Ri,j} [P_j^+ + P_j^-] \right] P_i^\pm, \quad C'_{Ri,j} = \begin{cases} C_R(\lambda_i, \lambda_j) & \text{if } \lambda_i > \lambda_j \\ \frac{\lambda_i}{\lambda_j} C_R(\lambda_j, \lambda_i) & \text{if } \lambda_i < \lambda_j. \end{cases}$$

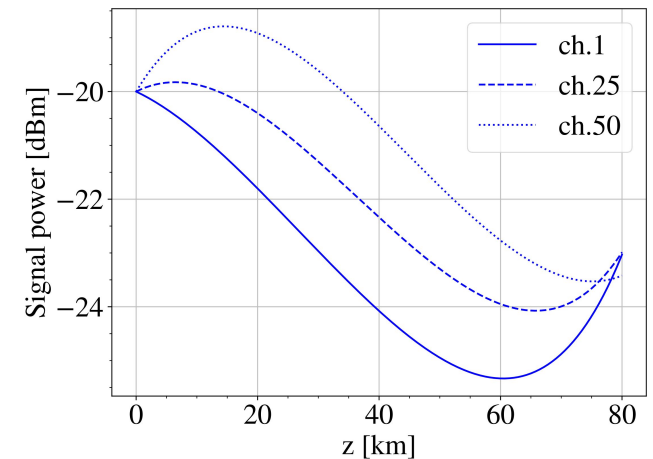
Results for -20dBm input power level, optimization for -3dB gain:



Copumping



Counterpumping



Bidirectional pumping

Coupled nonlinear Schrödinger equation model

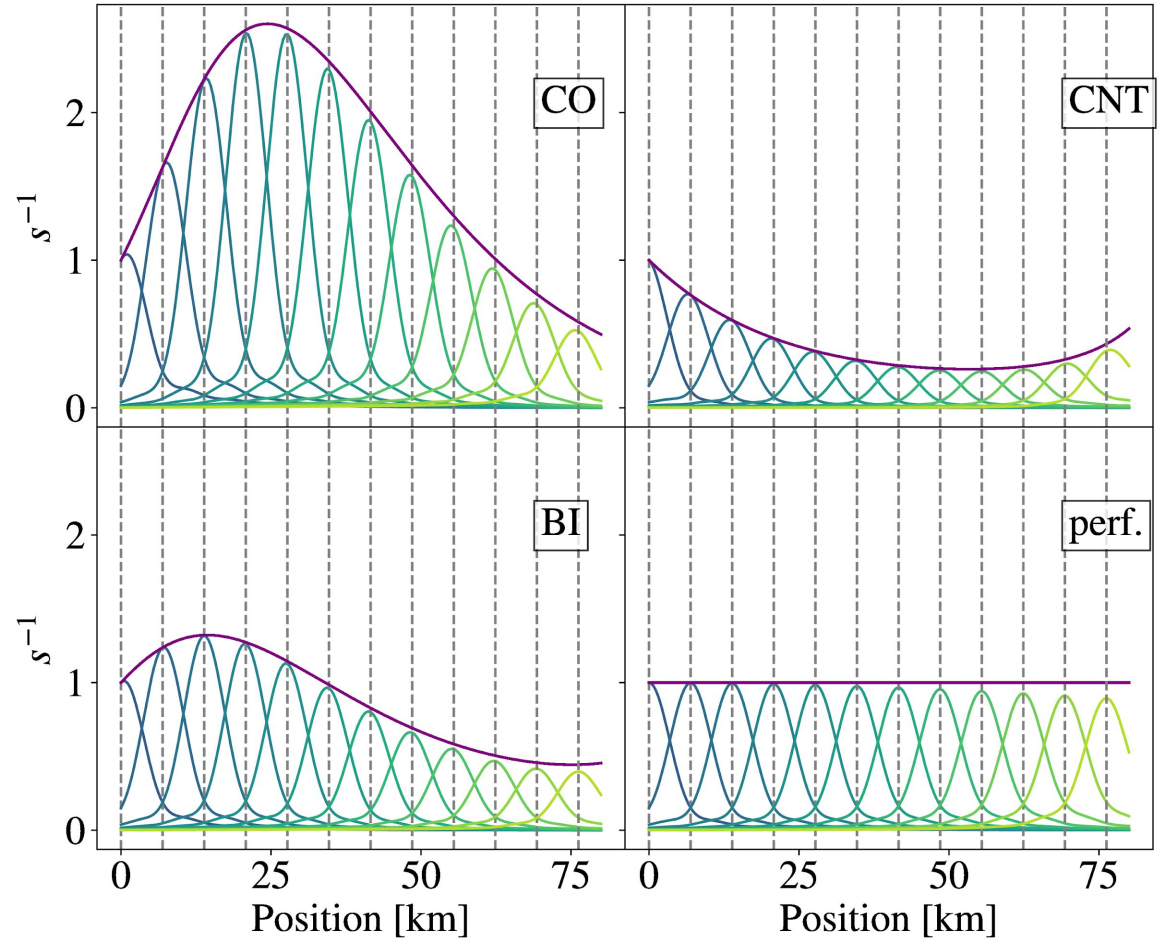
Channel dependence of Raman gain and attenuation is encoded in a coupled equation model:

$$\begin{aligned} \frac{\partial}{\partial z} u_A &= -i \frac{\beta_2}{2} \frac{\partial^2}{\partial t^2} u_A + \\ &+ i \gamma \left(f_A(z) |u_A|^2 + 2 f_B(z) |u_B|^2 \right) u_A \\ \frac{\partial}{\partial z} u_B &= -\Delta \beta_1 \frac{\partial}{\partial t} u_B - i \frac{\beta_2}{2} \frac{\partial^2}{\partial t^2} u_B + \\ &+ i \gamma \left(f_B(z) |u_B|^2 + 2 f_A(z) |u_A|^2 \right) u_B \end{aligned}$$

The **perturbation method** can be used in an analogous way, obtaining

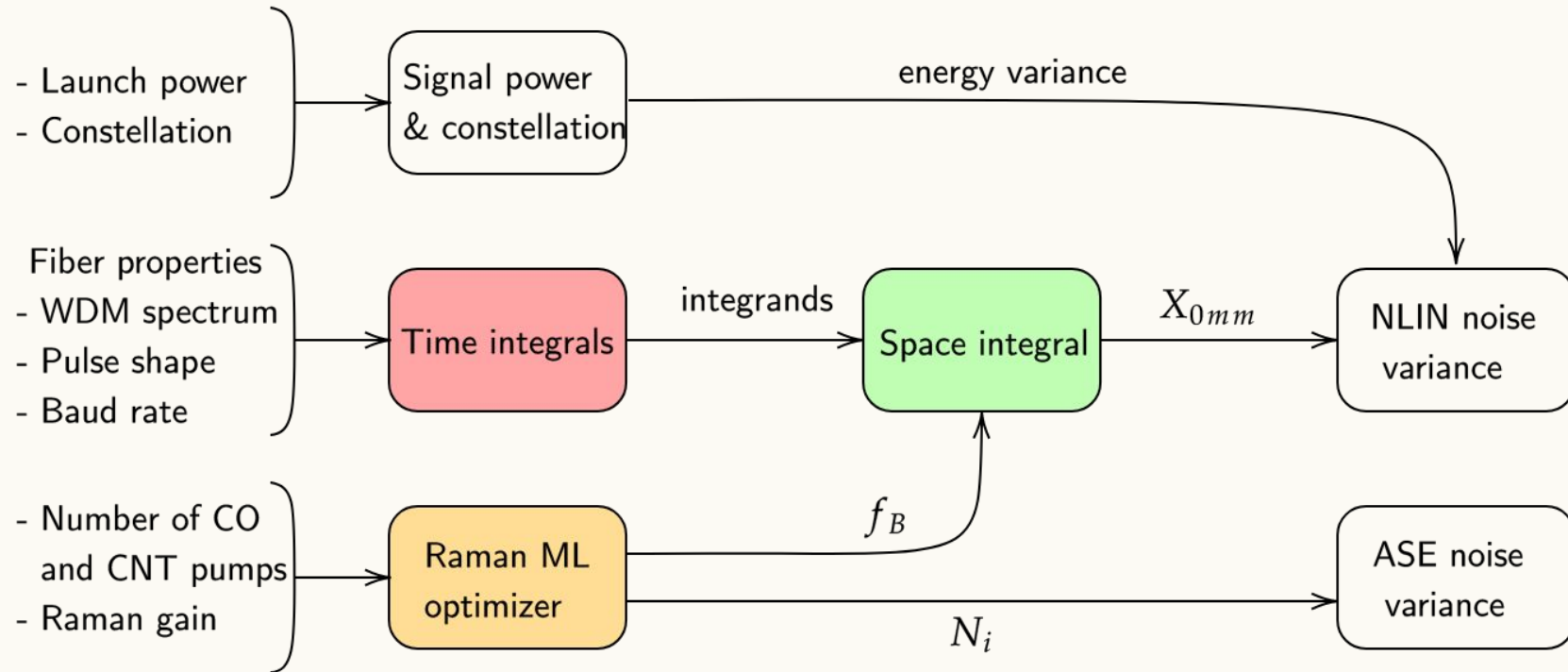
$$\begin{aligned} X_{0,m,m} &= \int_0^L dz f_B(z) \int_{-\infty}^{+\infty} dt \times \\ &\times \left| g^{(0)}(z, t) \right|^2 \left| g^{(0)}(z, t - mT - \beta_2 \Omega z) \right|^2 \end{aligned}$$

However, in this case, high dispersion approximation can **not** be used.



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Numerical simulations



Computation of the collisions

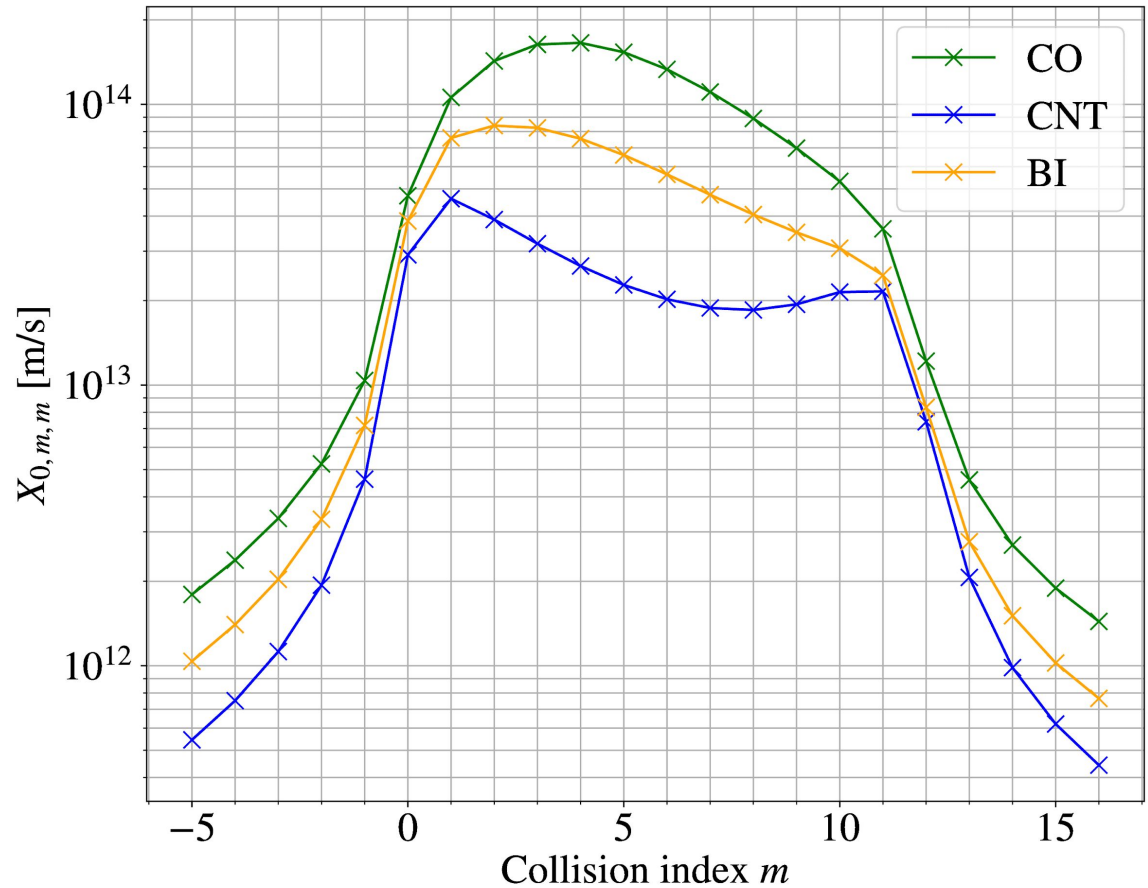
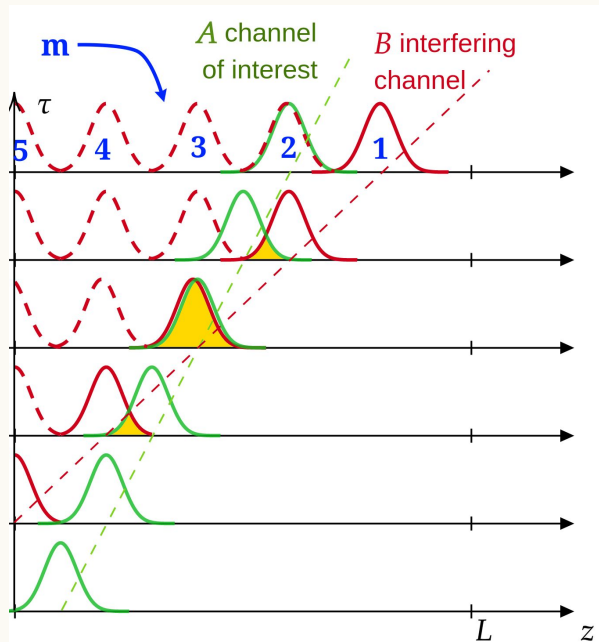
Computing the coefficient for each collision, **differences between pumping schemes** are evident.

Channels are 49 (interfering) and 50, with -20dBm input power.

CO: 8 copropagating pumps,

CNT: 10 counterpropagating pumps,

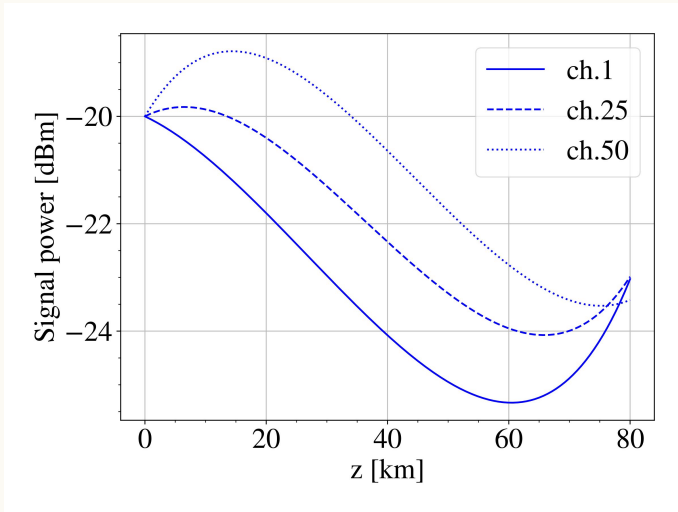
BI: 8 co-, 2 counterpropagating pumps.



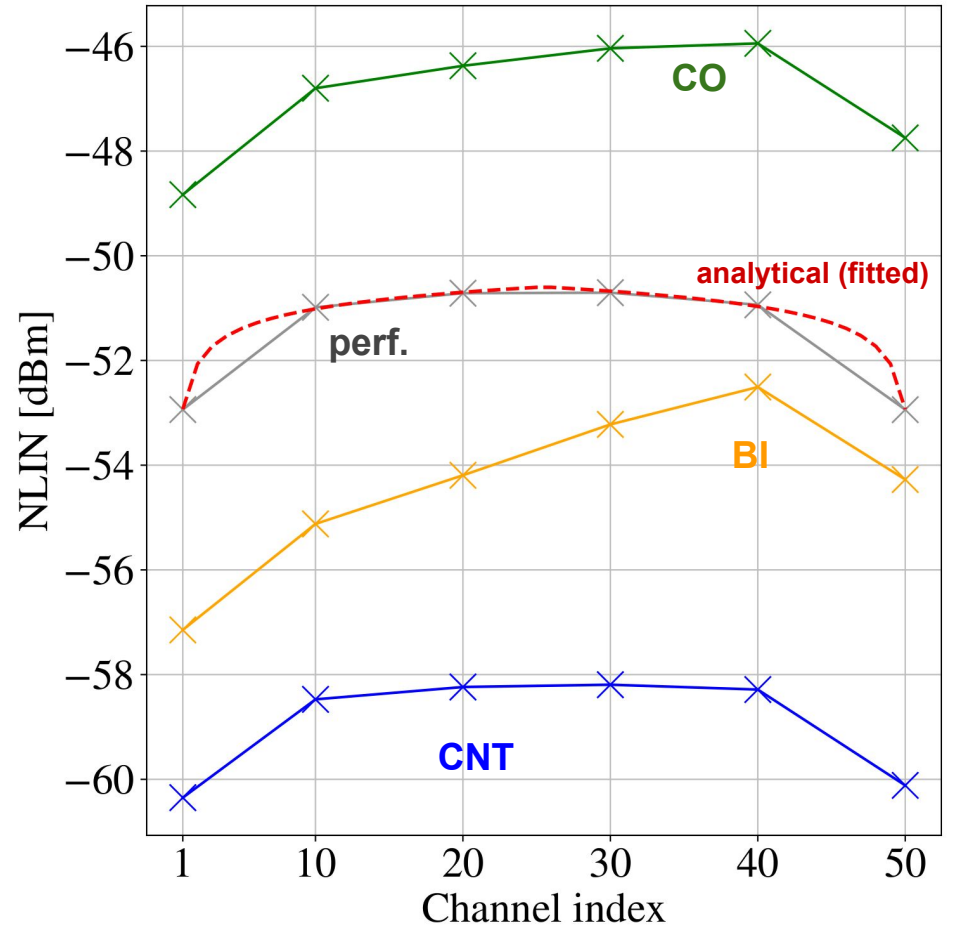
Impact of the position of the channel of interest

Asymmetry is due to different signal **power evolution profiles**.

This is particularly visible for the bidirectional case.



The **perfect amplification** profile is reported for comparison.



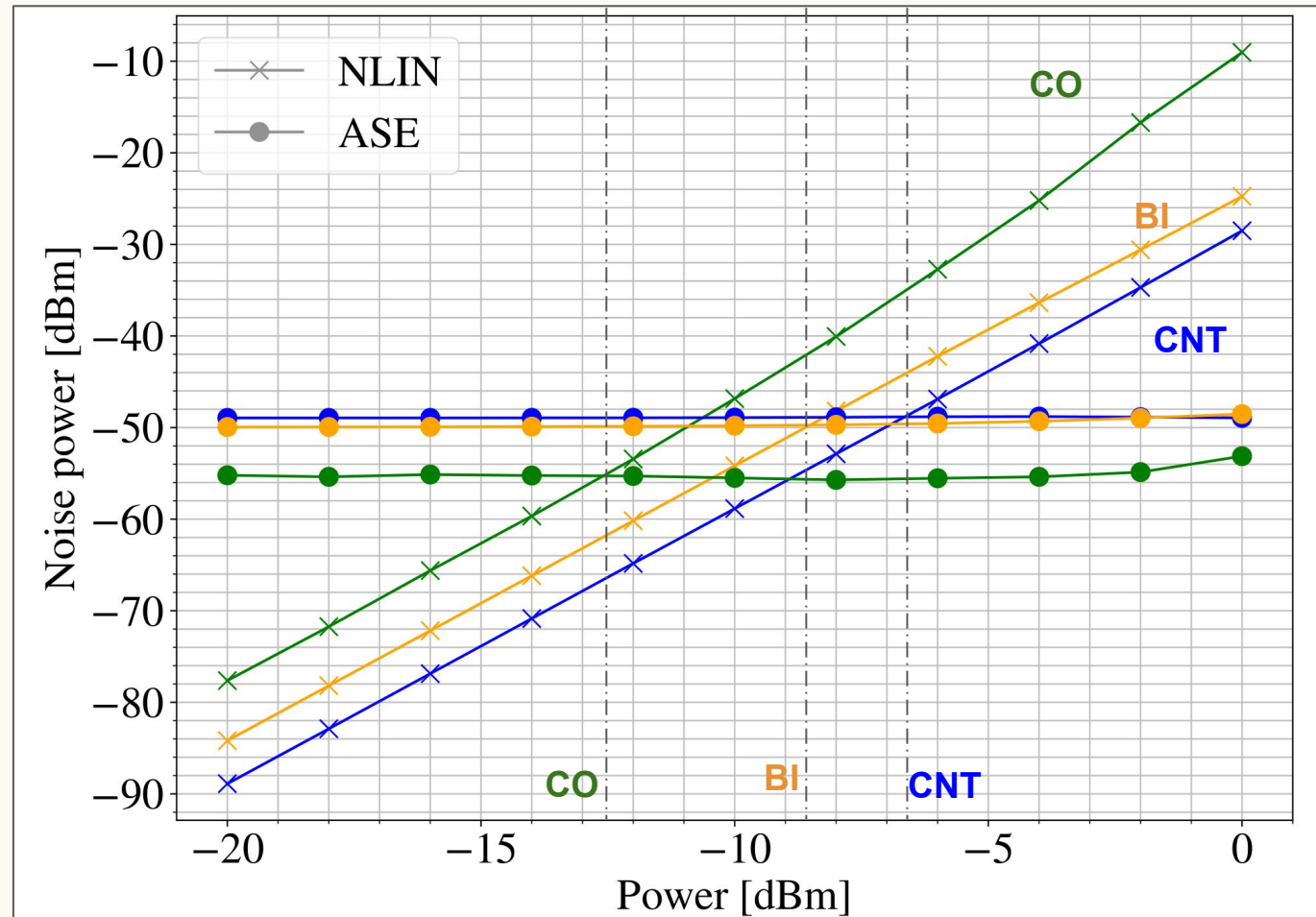
Impact of the launch power

ASE performance is similar in the various pumping schemes.

It is the dominant noise in the **low power regime**.

The **crossover** power level between regimes increases by preferring counterpropagating pumps.

Plot is the average of 6 equally spaced channels of interest.



Conclusions

In this work we studied

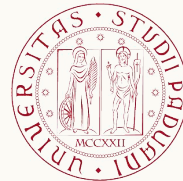
- How to generalize the **idealized model** for NLIN to **Raman-amplified WDM links**.
- The integration of the pumping scheme with **optimized equalization** into NLIN computations.
- The dependency of NLIN from **input power** and WDM **channel position**, and its comparison with ASE.

Further work may involve

- Study of NLIN dependence from **link length**, and comparison with ASE in this scenario.
- Design of an **optimization scheme** able to tackle not only flatness of the gain, but also **noise** performance.

Thanks for the attention

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References

- [1] J. Bromage, «Raman amplification for fiber communications systems», *Journal of Lightwave Technology*, vol. 22, n. 1, pagg. 79–93, 2004.
- [2] G. Marcon, A. Galtarossa, L. Palmieri, e M. Santagiustina, «Model-Aware Deep Learning Method for Raman Amplification in Few-Mode Fibers», *Journal of Lightwave Technology*, vol. 39, n. 5, pagg. 1371–1380, mar. 2021.
- [3] R. Dar, M. Feder, A. Mecozzi, e M. Shtaif, «Properties of nonlinear noise in long, dispersion-uncompensated fiber links», *Optics Express*, vol. 21, n. 22, pag. 25685, ott. 2013.
- [4] R. Dar, M. Feder, A. Mecozzi, e M. Shtaif, «Accumulation of nonlinear interference noise in fiber-optic systems», *Optics Express, OE*, vol. 22, n. 12, pagg. 14199–14211, giu. 2014.